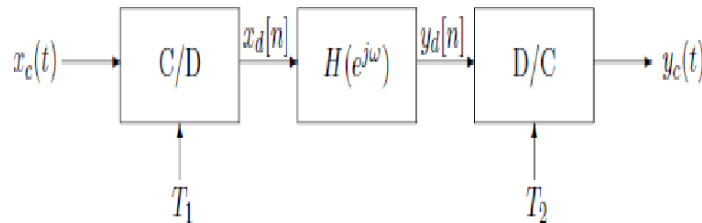


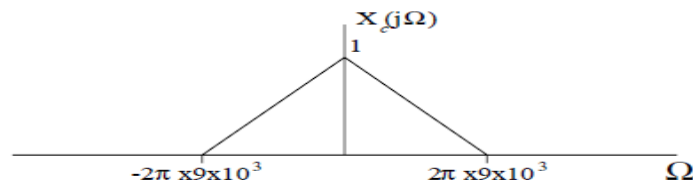
**Q.2** a. Define Quantization. Derive the signal-to-quantization noise ratio for sinusoidal signals.

**Answer:** Topic 4.8.3, Page Number 219 of Text Book 1

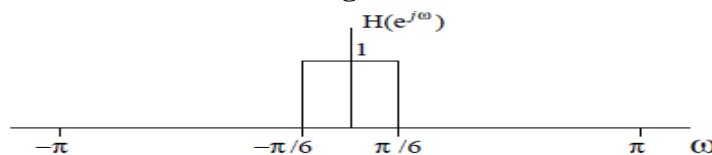
b. In the system shown in Fig.1,  $X_c(j\Omega)$  and  $H(e^{j\omega})$  are as shown and  $1/T_1 = 30000$ ,  $1/T_2 = 10000$  respectively. Sketch and label the Fourier transforms of  $y_d[n]$  and  $y_c(t)$ .



**Fig.1**

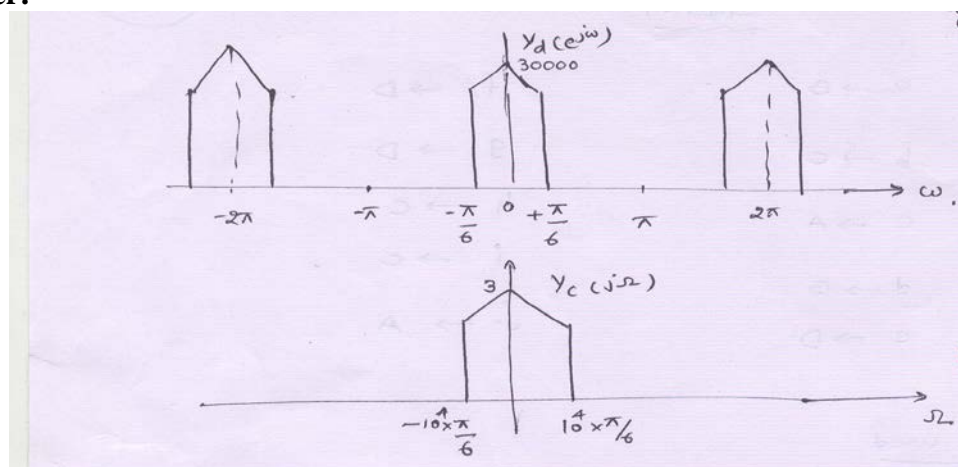


**Fig.2**



**Fig.3**

**Answer:**



- Q.3** a. Consider the LTI system with input  $x[n]$  and output  $y[n]$ , which are related through the difference equation:  $y[n] - \frac{5}{2}y[n-1] + y[n-2] = x[n]$
- Obtain the system function and its ROC
  - Draw its pole-zero plot
  - Comment on the causality and stability of this system

**Answer:**

**Q.3**  
 (a)  $y[n] - \frac{5}{2}y[n-1] + y[n-2] = x[n]$

(i) obtain system fun & its ROC (ii)

$\therefore H(z) = \frac{Y(z)}{X(z)} = \frac{1}{1 - \frac{5}{2}z^{-1} + z^{-2}}$

$= \frac{1}{(1 - \frac{1}{2}z^{-1})(1 - 2z^{-1})}$

Pole-zero plot.

(iii) ROC: three choices

If system, assumed, causal  $\rightarrow$  ROC is outside the outmost pole  
 and unstable. i.e.  $|z| > 2$

If system is, stable (assume)  $\rightarrow$  ROC will be  $\frac{1}{2} < |z| < 2$   
 and noncausal

For ROC  $|z| < \frac{1}{2}$ , system will be neither stable or causal,  
 anticausal and unstable.

- b. A discrete-time causal LTI system has the system function

$$H(z) = \frac{(1 + 0.2z^{-1})(1 - 9z^{-2})}{(1 + 0.81z^{-2})}$$

Find expression for a minimum-phase system  $H_1(z)$  and an all-pass system  $H_{ap}(z)$  such that  $H(z) = H_1(z) H_{ap}(z)$ .

**Answer:**

Q3(b)

$$H(z) = \frac{(1+0.2z^{-1})(1-9z^{-2})}{(1+0.81z^{-2})} \quad |26/9$$

$$= \left[ \frac{(1+0.2z^{-1})(1-\frac{1}{3}z^{-1})(1+\frac{1}{3}z^{-1})}{(1+0.81z^{-2})} \right] \left[ \frac{(1-\frac{1}{3}z^{-1})(1+\frac{1}{3}z^{-1})}{(1-\frac{1}{3}z^{-1})(1+\frac{1}{3}z^{-1})} \right]$$

$$= \left[ \frac{(1+0.2z^{-1})(1-\frac{1}{3}z^{-1})(1+\frac{1}{3}z^{-1})}{(1+0.81z^{-2})} \right] \left[ \frac{(1-\frac{1}{3}z^{-1})(1+\frac{1}{3}z^{-1})}{(1-\frac{1}{3}z^{-1})(1+\frac{1}{3}z^{-1})} \right]$$

$$H(z) = \left[ \frac{(1+0.2z^{-1})(1-\frac{1}{9}z^{-2})}{(1+0.81z^{-2})} \right] \left[ \frac{1-9z^{-2}}{1-\frac{1}{9}z^{-2}} \right]$$

3b.

where  $H_1(z) = \frac{(1+0.2z^{-1})(1-\frac{1}{9}z^{-2})}{(1+0.81z^{-2})}$  ... minimum-phase system

$H_2(z) = \frac{1-9z^{-2}}{1-\frac{1}{9}z^{-2}}$  ... all-pass system.

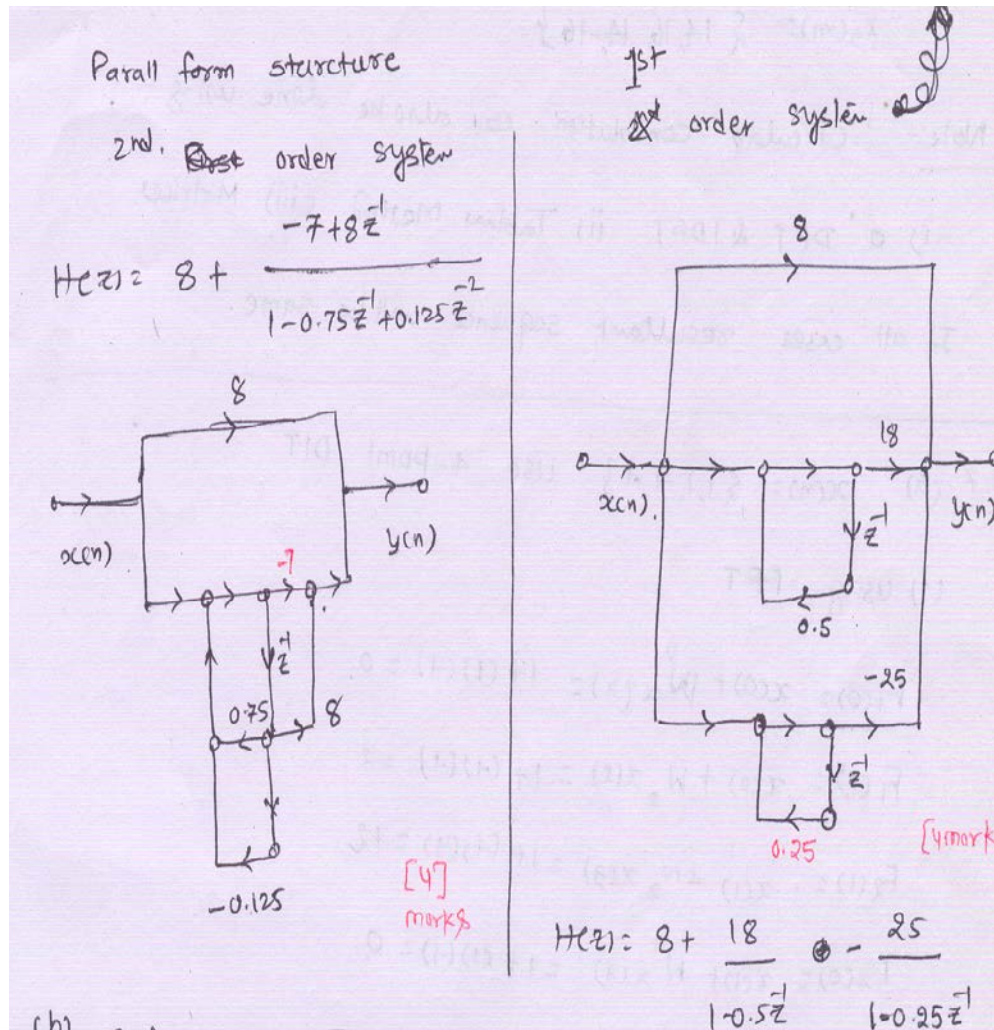
[08 mar]

- Q.4** a. Obtain the parallel-form structure of the given  $H(z)$  for first-order and second order systems.

$$H(z) = \frac{(1+2z^{-1}+z^{-2})}{(1-0.75z^{-1}+0.125z^{-2})}$$

**Answer:**





- b. Describe the signal flow graph representation of linear constant coefficient difference equations.

**Answer:** Topic 6.2 of Text Book 1

- Q.5** a. With an example, design a differentiator using Kaiser Window concept.

**Answer:** Topic 7.3.2 of Text Book 1

- b. Discuss the Parks- McClellan algorithm for type I low pass filter.

**Answer:** Topic 7.4.3 of Text Book 1

- Q.6** a. Discuss and prove the following properties of Discrete Fourier Transform.  
(i) Duality (ii) Symmetry

**Answer:** Topic 8.6.3 and 8.6.4 of Text Book 1

- b. Perform the Circular Convolution of the two sequences  $x_1(n) = \{2, 1, 2, 1\}$  and  $x_2(n) = \{1, 2, 3, 4\}$ .

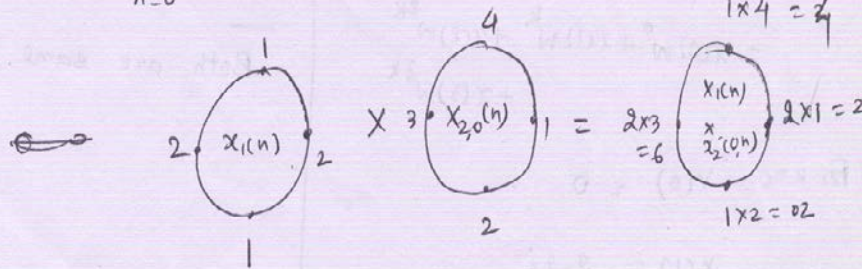
**Answer:**

(b) circular convolution

$$x_1(n) = \{2, 1, 2, 1\} \quad \& \quad x_2(n) = \{1, 2, 3, 4\}$$

$$x_3(m) = \sum_{n=0}^{N-1} x_1(n) \cdot x_2(m-n) = \sum_{n=0}^{N-1} x_1(n) \cdot x_{2,m}(n)$$

$$m=0, \quad x_3(0) = \sum_{n=0}^3 x_1(n) \cdot x_{2,0}(n)$$



$$\therefore x_3(0) = 2 + 2 + 6 + 4 = 14$$

4/6

similarly

$$m=1 \quad x_3(1) = 16$$

$$m=2 \quad x_3(2) = 14$$

$$m=3 \quad x_3(3) = 16$$

$$\therefore x_3(m) = \{14, 16, 14, 16\}$$

Note:- Circular convolution can also be done using

- i) a DFT & IDFT    ii) Tables Method    (iii) Matrices.

In all cases resultant sequence will be same.

- Q.7 a. For  $x(n) = (1, 1, -1, -1)$  use 4-point DIT, algorithm for FFT and cross check the result using DFT.



Answer:

Q:7 (a)  $x(n) = \{1, 1, -1, -1\}$  use 4-point DIT

(i) using FFT

$$F_1(0) = x(0) + W_4^0 x(2) = 1 + (1)(-1) = 0$$

$$F_1(1) = x(0) + W_4^1 x(2) = 1 + (-1)(-1) = 2$$

$$F_2(1) = x(1) + W_4^2 x(3) = 1 + (1)(-1) = 0$$

$$F_2(0) = x(1) + W_4^0 x(3) = 1 + (1)(-1) = 0$$

$$\therefore X(0) = F_1(0) + W_4^0 F_2(0) = 0 + (1)(0) = 0$$

$$X(1) = F_1(1) + W_4^1 F_2(1) = 2 + (j)(0) = 2 - 2j$$

$$X(2) = F_1(0) + W_4^2 F_2(0) = 0 + (-1)(0) = 0$$

$$X(3) = F_1(1) + W_4^3 F_2(1) = 2 + j(0) = 2 + 2j$$

$\therefore X(k) = \{0, 2 - 2j, 0, 2 + 2j\}$  [4 marks]

Cross check using DFT

$$X(k) = \sum_{n=0}^3 x(n) W_4^{nk}$$

$$= x(0)W_4^{0k} + x(1)W_4^{1k} + x(2)W_4^{2k} + x(3)W_4^{3k}$$

For  $k=0$ ,  $X(0) = 0$

$$X(1) = 2 - 2j$$

$$X(2) = 0$$

$$X(3) = 2 + 2j$$

hence  $X(k) = \{0, 2 - 2j, 0, 2 + 2j\}$

Both are same. [4 marks]

**Q.8** a. Discuss the effect of windowing on Fourier analysis of sinusoidal signals.

**Answer:** Section 10.2, Page Number 723 of Text Book 1

b. Discuss the time-dependent Fourier transform with a suitable example.

**Answer:** Topic 10.3 of Text Book 1

- Q.9** b. For a real, causal sequence  $x(n]$  for which  $X_R(e^{j\omega}) = \frac{5}{4} - \cos \omega$ . Obtain
- The original sequence  $x(n]$  and
  - Imaginary part of the Fourier transform  $X_I(e^{j\omega})$ .

**Answer:**

Q.9(b) (i)  $X_R(e^{j\omega}) = \frac{5}{4} - \cos \omega = \frac{5}{4} - \frac{1}{2}e^{j\omega} - \frac{1}{2}e^{-j\omega}$

we know that  $x_e(n) \longleftrightarrow X_R(e^{j\omega})$

$$x_e(n) = \text{IDTFT}[X_R(e^{j\omega})]$$

$$x_e(n) = \text{IDTFT}\left[\frac{5}{4} - \frac{1}{2}e^{j\omega} - \frac{1}{2}e^{-j\omega}\right]$$

$$x_e(n) = \frac{5}{4}\delta(n) - \frac{1}{2}\delta(n+1) - \frac{1}{2}\delta(n-1)$$

$$\therefore x(n) = 2x_e(n)u(n) - x_e(0)\delta(n)$$

$$x(n) = \left[\frac{5}{2}\delta(n) - \delta(n-1)\right] - \frac{5}{4}\delta(n)$$

$$x(n) = \frac{5}{4}\delta(n) - \delta(n-1)$$

$$X(e^{j\omega}) = \frac{5}{4} - e^{-j\omega}$$

$$= \left[\frac{5}{4} - \cos \omega\right] + j \sin \omega$$

$$X(e^{j\omega}) = X_R(e^{j\omega}) + j X_I(e^{j\omega})$$

$$\therefore X_I(e^{j\omega}) = \sin \omega$$

### TEXT BOOKS

Discrete-Time Signal Processing (1999), Oppenheim, A. V., and Schaffer, R. W., with J. II R. Buck, Second Edition, Pearson Education, Low Price Edition.